# PAPER Analysis of Phase-Inversion Waves in Coupled Oscillators Synchronizing at In-and-Anti-Phase

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**SUMMARY** Recently, we have discovered wave propagation phenomena which are continuously existing waves of changing phase states between two adjacent oscillators from in-phase to anti-phase or from anti-phase to in-phase in van der Pol oscillators coupled by inductors as a ladder. We named the phenomena as "phase-inversion waves." In this study, phase-inversion waves which exist in the state of in-and-anti-phase synchronization have been found. We observe the phenomena by circuit experiments and computer calculations, and investigate them.

key words: coupled oscillators, phase difference, phaseinversion waves, in-and-anti-phase synchronization

# 1. Introduction

Many studies on synchronization phenomena of coupled oscillators have been carried out up to now in various fields, physics [1]–[3], biology [4], electrical engineering [5]–[14], and so on. Endo et al. have reported details of theoretical analysis and circuit experiments about some coupled oscillators as a ladder, a ring and a twodimensional array [6]–[8].

Recently, the authors have discovered a wave propagation phenomena that phase states between adjacent oscillators change from in-phase to anti-phase or from anti-phase to in-phase in oscillators coupled by inductors as a ladder [11]. We named the phenomena as "phase-inversion waves." It is very important to analyze the phase-inversion waves and to clarify the mechanism of the generation, because it is similar to propagation phenomena of electrical information in an axial fiber of nervous system. In [11], we explained the mechanism of the generation of the phase-inversion waves by using the relationship between phase states and instantaneous oscillation frequencies. Further, we confirmed that the phenomena could be explained by a simplified mathematical model [12].

In this study, we investigate the phase-inversion waves observed from the same van der Pol oscilla-

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tors ladder but in different synchronization mode; inand-anti phase synchronization, which is similar to the phenomenon reported in [10]. In this synchronization mode, in-phase and anti-phase synchronizations exist alternately. An interesting point of this synchronization is that the edge oscillators and their adjacent oscillators must be synchronized at anti-phase. This causes various limitations of the generated phenomena with respect to the phase-inversion waves. For example, in the past studies, only even numbers of the phase-inversion waves can exist and the number of the oscillators do not affect the observed phenomena. On the other hand, in this study, even numbers of the phase-inversion waves exist in even numbers of the oscillators and odd numbers of the phase-inversion waves exist in odd numbers of the oscillators. The reason of this difference is explained in Sect. 4. Further, the mechanisms of propagation, reflection at an edge of the array and reflection by collision of two phase-inversion waves are different from those reported in [11]. They are explained by using a relationship between phase states and instantaneous oscillation frequencies in Sect. 5. The computer simulated results are confirmed by circuit experiments.

# 2. Circuit Model

The circuit model used in this study is shown in Fig. 1. N van der Pol oscillators are coupled by inductors  $L_0$ . We carried out computer calculations for the cases of N = 9 and 20. Further, we carried out circuit experiments for the case of N = 9. In the computer calculations



 $Fig. 1 \quad {\rm Coupled \ van \ der \ Pol \ oscillators \ as \ a \ ladder}.$ 

lations, we assume the v - i characteristics of the nonlinear negative resistors in the circuit by the following function.

$$i_r(v_k) = -g_1 v_k + g_3 v_k^3 \quad (g_1, g_3 > 0).$$
(1)

The circuit equations governing the circuit in Fig. 1 are written as

[First Oscillator]

$$\dot{x}_1 = y_1$$

$$\dot{y}_1 = -x_1 + \alpha (x_2 - x_1) + \varepsilon \left( y_1 - \frac{1}{3} y_1^3 \right)$$
(2)

[Middle Oscillators]

$$\dot{x}_{k} = y_{k}$$

$$\dot{y}_{k} = -x_{k} + \alpha(x_{k+1} - 2x_{k} + x_{k-1}) + \varepsilon \left(y_{k} - \frac{1}{3}y_{k}^{3}\right)$$

$$(k = 2 \sim N-1)$$
(3)

[Last Oscillator]

$$\dot{x}_N = y_N \tag{4}$$
$$\dot{y}_N = -x_N + \alpha (x_{N-1} - x_N) + \varepsilon \left( y_N - \frac{1}{3} y_N^3 \right),$$

where

$$t = \sqrt{L_1 C} \tau, \ i_{L_1 k} = \sqrt{\frac{Cg_1}{3L_1 g_3}} x_k, v_k = \sqrt{\frac{g_1}{3g_3}} y_k,$$
  
$$\alpha = \frac{L_1}{L_0}, \ \varepsilon = g_1 \sqrt{\frac{L_1}{C}}, \ \frac{d}{d\tau} = "\cdot".$$
(5)

It should be noted that  $\alpha$  corresponds to the coupling of the oscillators and  $\varepsilon$  corresponds to the nonlinearity of the oscillators. Throughout the paper, we fix  $\alpha = 0.10$  and  $\varepsilon = 0.30$  and calculate (2)–(4) by using the fourth-order Runge-Kutta method with the stepsize  $\Delta \tau = 0.01$ .

#### 3. Example of Phase-Inversion Waves

Figure 2 shows some typical examples of phaseinversion waves reported previously. They are computer calculated results from the circuit with the sizes of N = 29 and 30. In the figures, the vertical axes are the sum of the voltages of adjacent oscillators and the horizontal axis is time. White regions in the diagram correspond to the states that the sum of the voltages is close to zero, namely the adjacent two oscillators are synchronized at anti-phase. While, black regions correspond to the states that the sum of the voltages has large amplitude. We can see that the adjacent two oscillators are synchronized at in-phase in the black regions from Fig. 3, which shows the same phenomenon as Fig. 2(b), but the vertical axes are the difference between the voltages of adjacent oscillators. Namely, we



(a) Double waves (N = 29).



(b) Double waves (N = 30).



(c) Extinction by collision of waves (N = 30).



(d) Reflection by collision of two waves (N = 30).





**Fig. 2** Examples of phase-inversion waves.  $y_k + y_{k+1}$  vs. time  $(k = 1 \sim N - 1)$ .  $\alpha = 0.10$ ,  $\varepsilon = 0.30$  and  $\Delta \tau = 0.01$ .



Double waves (N = 30).

**Fig. 3** Examples of phase-inversion waves.  $y_k - y_{k+1}$  vs. time  $(k = 1 \sim N - 1)$ .  $\alpha = 0.10$ ,  $\varepsilon = 0.30$  and  $\Delta \tau = 0.01$ .

can say that the phase-inversion waves in Fig. 2 propagate in the state of in-phase synchronization.

#### Remark A:

Strictly speaking, when the phase-inversion waves propagate, the whole circuits are not synchronized. However, because the concept of synchronization is useful to explain the phenomena, the phase-inversion waves are described such as they propagate in a synchronization state.

In Figs. 2 and 3, we can see the wave propagation, reflection and extinction. Even numbers of the phase-inversion waves exist in both odd and even numbers of coupled oscillators.

It has been known that two van der Pol oscillators coupled by an inductor have two stable phase states, namely in-phase synchronization and anti-phase synchronization. Further, the oscillation frequency for the in-phase synchronization is smaller than the oscillation frequency for the anti-phase synchronization. By using these features, we could explain the mechanism of the phenomena in Fig. 2 by using the relationship between phase states and instantaneous oscillation frequencies (See [11] for the details).

# 4. Phase-Inversion Waves in the State of Inand-Anti-Phase Synchronization

#### 4.1 In-and-Anti-Phase Synchronization

In the circuit model, we can observe another type of synchronization as shown in Fig. 4. The vertical axes are the amplitudes of the voltages and the horizontal axis is time. We name this synchronization state as "in-and-anti-phase synchronization" because in-phase and anti-phase exist alternately. Also, in this synchronization state, the edge oscillators and their adjacent oscillators can not be synchronized at in-phase. Hence, this synchronization state can be observed only when N is an even number. The oscillation frequency for the in-and-anti-phase synchronization  $f_{mid}$  is almost the average of the oscillation frequency for the in-phase synchronization frequency for the in-phase synchronizatio



Fig. 4 Example of in-and-anti-phase synchronization. (a) Circuit experimental result for  $L_0=1170$  mH,  $L_1=200$  mH, C=68 nF and r=1 k $\Omega$ . (b) Computer calculated result for  $\alpha = 0.10$ ,  $\varepsilon = 0.30$  and  $\Delta \tau = 0.01$ .



**Fig. 5** Experimental results. Circuit experiments:  $L_0 = 1170 \text{ mH}$ ,  $L_1 = 200 \text{ mH}$ , C = 68 nF and  $r=1 \text{ k}\Omega$ . Computer calculations:  $\alpha = 0.10$ ,  $\varepsilon = 0.30$  and  $\Delta \tau = 0.01$ .

the anti-phase synchronization  $f_{high}$ .

### 4.2 Phase-Inversion Waves

Figures 5 and 6 show examples of phase-inversion waves in the state of in-and-anti-phase synchronization. The vertical axes of Figs. 5 and 6 are the sum of the voltages of adjacent oscillators and the horizontal axis is time.





(b) Triple waves (N = 29).



(c) Double waves (N = 30).



(d) Double waves (N = 30).



Fig. 6 Examples of phase-inversion waves.  $\alpha = 0.10$ ,  $\varepsilon = 0.30$  and  $\Delta \tau = 0.01$ .

Figure 5 shows circuit experimental results and the corresponding computer calculated results obtained for the cases of  $N \leq 9$ . In Fig. 5(a), phase-inversion waves do not exist. A phase-inversion-wave exists continuously in Fig. 5(b). We can see that the change of the phase states from in-phase to anti-phase or from anti-phase to in-phase propagates. In Fig. 5(c), a pair of phase-inversion-waves exist.

Figure 6 shows computer calculated results obtained for the cases of N = 29 and 30. We can observe phase-inversion waves similar to Fig. 5.

In in-and-anti-phase synchronization, we could not observe wave extinction by collision of two waves like Fig. 2(c).

## 4.3 Single Phase-Inversion Wave

In this subsection, We clarify the relationship between the number of oscillators and the number of phaseinversion waves.

<in In-Phase Synchronization>

In the state of the in-phase synchronization, the edge oscillators and their adjacent oscillators must be synchronized at in-phase. Therefore, in spite of the number of the oscillators in the array, only even numbers of phase-inversion waves can exist [11],

<in In-and-Anti-Phase Synchronization>

In the state of the in-and-anti-phase synchronizing, the edge oscillators and their adjacent oscillators must be synchronized at anti-phase. Therefore, when an even number of oscillators are coupled, even numbers of phase-inversion waves can exist, while when an odd number of oscillators are coupled, odd numbers of phase-inversion waves can exist.

#### Remark B:

As explained in Sect. 4.1, the in-and-anti-phase synchronization is not stable in the array of odd numbers of oscillators. However, when odd numbers of phaseinversion waves exist in the array, the in-and-anti-phase synchronization can be observed. Furthermore, for the first time, odd numbers of phase-inversion waves including the single wave are observed.

# 5. Mechanisms of Phase-Inversion Waves

Figures 7, 8 and 9 show phase differences and instantaneous frequencies, where  $\Phi_{k,k+1}$  is phase difference between OSC<sub>k</sub> and OSC<sub>k+1</sub> and  $f_k$  is instantaneous frequency of OSC<sub>k</sub>. We define them as follows:

$$\Phi_{k,k+1} = \frac{\tau_k(n) - \tau_{k+1}(n)}{\tau_k(n) - \tau_k(n-1)} \times \pi$$
(6)

$$f_k(n) = \frac{1}{2(\tau_k(n) - \tau_k(n-1))}$$
(7)

where  $\tau_k(n)$  is time when the voltage of OSC<sub>k</sub> crosses



Fig. 7 Mechanism of propagation (computer calculated results).



Fig. 8 Mechanism of reflection at an edge of the array (computer calculated results).



Fig. 9 Mechanism of reflection by collision of two waves (computer calculated results).

0 [V] at *n*-th time. It has been known that f changes as  $\Phi$  changes. We explain the mechanisms of the phase-inversion waves by using the relationship between f and  $\Phi$ .

# 5.1 Mechanism of Propagation

Figure 7 shows propagating single phase-inversion wave. The mechanism of the wave propagation can be explained in Table 1.

After the wave reflects at an edge of the array, when the propagating wave reaches  $OSC_4$  from the direction of  $OSC_N$ , phase states and instantaneous frequencies change in a similar manner (see  $\otimes$  in Fig. 7(a) and (b)).

5.2 Mechanism of Reflection at an Edge

Figure 8 shows reflecting single phase-inversion wave at an edge of the array. The mechanism of the reflection at an edge of the array can be explained in Table 2.

5.3 Mechanism of Reflection by Collision

Figure 9 shows reflection by collision of two phaseinversion waves in the middle of the array. When the two phase-inversion waves reach two adjacent oscillators at the same time, the phase-inversion waves reflect. The mechanism of the wave reflection can be explained

time $[\tau]$	sequence of propagation	
Let us assume that the circuit synchronizes at in-and-anti- phase and that single wave reaches $OSC_6$ from the direction of $OSC_N$ .		
around 24	$f_6$ changes from $f_{mid}$ toward $f_{high}$ , because the phase state between OSC <sub>6</sub> and OSC <sub>7</sub> changes from in-phase to anti-phase (see ① in Fig. 7(b)).	
around 27	The phase state of OSC <sub>6</sub> advances, because $f_6$ changes from $f_{mid}$ toward $f_{high}$ . Hence, $\Phi_{5,6}$ starts to change from $-\pi$ to 0 (see Eq. (6) and 2) in Fig. 7(a)).	
around 32	$f_5$ starts to change from $f_{mid}$ toward $f_{low}$ (see ③ in Fig. 7(b)).	
around 36	$\Phi_{4,5}$ changes from 0 to $-\pi$ (see $\textcircled{4}$ in Fig. 7(a)).	
around 62	$f_6$ changes to $f_{mid}$ again before $f_6$ reaches $f_{high}$ , because the phase states between OSC <sub>5</sub> and OSC <sub>6</sub> and between OSC <sub>6</sub> and OSC <sub>7</sub> start to in- terchange. (see $\textcircled{5}$ in Fig. 7(b)).	
around 79	$f_5$ changes to $f_{mid}$ again before $f_5$ reaches $f_{low}$ because the phase states between OSC <sub>5</sub> and OSC <sub>6</sub> and between OSC <sub>6</sub> and OSC <sub>7</sub> start to interchange (see (6) in Fig. 7(b)).	
around 114	$f_5$ becomes to $f_{mid}$ when the phase states be- tween OSC <sub>4</sub> and OSC <sub>5</sub> and between OSC <sub>5</sub> and OSC <sub>6</sub> finish to interchange (see $\bigcirc$ in Fig. 7(a) and (b)).	

in Table 3.

# 5.4 Extinction by Collision

We could observe extinction of the phase-inversion waves by collision in the past study [11] (see Fig. 2(c)). However, such an extinction can not be observed in this study. The reason would be explained as follows. In the past study, when two phase-inversion waves reach an oscillator at the same time, the phase-inversion waves disappeare. Because the instantaneous frequency of the oscillator changes from  $f_{low}$  to  $f_{high}$  or  $f_{high}$  to  $f_{low}$ . However, in this study, the instantaneous frequency of the oscillator can not change as the past study, because the changing directions of the instantaneous frequency are different between the two sides of the oscillator, for example, from  $f_{mid}$  to  $f_{high}$  at the right side and from  $f_{mid}$  to  $f_{low}$  at the left side. Therefore, we can not

time $[\tau]$	sequence of reflection at an edge	
Let us assume that the circuit synchronizes at in-and-anti- phase and that single wave reaches $OSC_3$ from the direction of $OSC_N$ . The phase state between $OSC_1$ and $OSC_2$ must be anti-phase and the phase state between $OSC_2$ and $OSC_3$ must be in-phase (see Sect. 4.1).		
around 80	$f_2$ starts to change from $f_{mid}$ toward $f_{high}$ , because the phase state between OSC <sub>2</sub> and OSC <sub>3</sub> changes from in-phase to anti-phase (see ① in Figs. 8(a) and (b)).	
around 84	$\Phi_{1,2}$ starts to change from $-\pi$ to 0 (see 2) in Fig. 8(b)).	
around 89	$f_1$ starts to degrease toward $f_{low}$ (see (3) in Fig. 8(a)).	
around 122	$f_2$ starts to decrease to $f_{mid}$ again before $f_2$ reaches $f_{high}$ , because the phase states between OSC <sub>1</sub> and OSC <sub>2</sub> and between OSC <sub>2</sub> and OSC <sub>3</sub> start to interchange (see $$ in Fig. 8(b)).	
around 148	When $\Phi_{1,2}$ reaches 0, $f_1$ reaches $f_{low}$ and $f_2$ reaches $f_{mid}$ . However, the in-phase synchro- nization of OSC <sub>1</sub> and OSC <sub>2</sub> is not stable, be- cause $f_1$ is not equal to $f_2$ . $\Phi_{1,2}$ changes to $\pi$ , because $f_1$ is smaller than $f_2$ . Further, $f_1$ and $f_2$ start to increase toward $f_{high}$ (see (5) in Figs. 8(a) and (b)).	
around 154	$\Phi_{2,3}$ starts to change to $-2\pi$ , because $f_2$ is larger than $f_3$ (see (6) in Fig. 8(a)).	
around 160	$f_3$ changes toward $f_{low}$ . (see $\bigcirc$ in Fig. 8(b)).	
around 170	$f_2$ changes to $f_{mid}$ again before $f_2$ reaches $f_{high}$ , because the phase states between OSC <sub>1</sub> and OSC <sub>2</sub> and between OSC <sub>2</sub> and OSC <sub>3</sub> start to in- terchange (see $\circledast$ in Fig. 8(b)).	
around 200	When $\Phi_{1,2}$ reaches $\pi$ , $f_1$ and $f_2$ reaches $f_{mid}$ again (see $\textcircled{9}$ in Figs. 8(a) and (b)).	

 Table 2
 Mechanism of reflection at an edge.

observe extinction by collision.

## 6. Conclusions

In this study, we observed phase-inversion waves in a ladder of coupled oscillators synchronizing at in-andanti-phase by both circuit experiments and computer calculations.

In the previous study, wave propagation, reflection

Table 3 Mechanism of reflection by collision

time $[\tau]$	sequence of reflection by collision	
Let us assume that the circuit synchronizes at in-and-anti- phase and that two phase-inversion waves are going to reach $OSC_{13}$ and $OSC_{16}$ at the same time from the direction of $OSC_1$ and $OSC_1$ , respectively. The phase states between $OSC_{13}$ and $OSC_{14}$ and between $OSC_{15}$ and $OSC_{16}$ must be anti-phase and the phase state between $OSC_{14}$ and $OSC_{15}$ must be in-phase (see Sect. 4.1).		
around 51	$f_{14}$ and $f_{15}$ start to change from $f_{mid}$ toward $f_{low}$ , because $\Phi_{13,14}$ changes from $-\pi$ to $-2\pi$ . $\Phi_{15,16}$ changes from $-\pi$ to 0 (see $\oplus$ in Fig. 8(b)).	
around 60	$\Phi_{14,15}$ does not change, because $f_{14}$ and $f_{15}$ start to change toward $f_{low}$ at the same time (see 2) in Fig. 8(a)).	
around 78	$f_{13}$ and $f_{16}$ start to decrease to $f_{mid}$ again be- fore they reach $f_{high}$ , because the phase states between OSC <sub>12</sub> and OSC <sub>13</sub> and between OSC <sub>13</sub> and OSC <sub>14</sub> start to interchange and the phase states between OSC <sub>15</sub> and OSC <sub>16</sub> and between OSC <sub>16</sub> and OSC <sub>17</sub> also start to interchange (see ③ in Fig. 8(b)).	
around 103	When $\Phi_{13,14}$ and $\Phi_{15,16}$ reach $-2\pi$ and 0, $f_{14}$ and $f_{15}$ reach $f_{low}$ . However, because $f_{13}$ and $f_{16}$ are $f_{mid}$ , $\Phi_{13,14}$ and $\Phi_{15,16}$ change to $-3\pi$ and $\pi$ . Therefore $f_{13}$ and $f_{16}$ change from $f_{mid}$ toward $f_{high}$ , and $f_{14}$ and $f_{15}$ change from $f_{low}$ to $f_{mid}$ (see ④ in Figs. 8(a) and (b)).	
around 300	When $\Phi_{13,14}$ and $\Phi_{15,16}$ reach $-3\pi$ and $\pi$ , $f_{13} \sim f_{16}$ reach $f_{mid}$ .	

at an edge of the array, reflection by collision of two waves and extinction by collision of two waves could be observed. However, the extinction could not be observed in the in-and-anti-phase synchronization. Further, single phase-inversion wave could be observed for the first time.

In the state of the in-and-anti-phase synchronization, instantaneous oscillation frequencies change around only single frequency  $f_{mid}$ . We could explain the mechanism of the wave propagation, the reflection by collision of two waves and the reflection at an edge of the array by using the relationship between phase states and instantaneous frequencies.

#### References

 L.L. Bonilla, C.J. Pérez Vicente, and R. Spigler, "Timeperiodic phases in populations of nonlinearly coupled oscillators with bimodal frequency distributions," Physica D: Nonlinear Phenomena, vol.113, no.1, pp.79–97, Feb. 1998.

- [2] J.A. Sherratt, "Invading wave fronts and their oscillatory wakes are linked by a modulated traveling phase resetting wave," Physica D: Nonlinear Phenomena, vol.117, no.1-4, pp.145–166, June 1998.
- [3] G. Abramson, V.M. Kenkre, and A.R. Bishop, "Analytic solutions for nonlinear waves in coupled reacting systems," Physica A: Statistical Mechanics and its Applications, vol.305, no.3-4, pp.427–436, March 2002.
- [4] C.M. Gray, "Synchronous oscillations in neuronal systems: Mechanisms and functions," J. Computational Neuroscience, vol.1, pp.11–38, 1994.
- [5] T. Suezaki and S. Mori, "Mutual synchronization of two oscillators," IECE Trans., vol.48, no.9, pp.1551–1557, Sept. 1965.
- [6] T. Endo and S. Mori, "Mode analysis of a multimode ladder oscillator," IEEE Trans. Circuits Syst., vol.23, no.2, pp.100–113, Feb. 1976.
- [7] T. Endo and S. Mori, "Mode analysis of two-dimensional low-pass multimode oscillator," IEEE Trans. Circuits Syst., vol.23, no.9, pp.517–530, Sept. 1976.
- [8] T. Endo and S. Mori, "Mode analysis of a ring of a large number of mutually coupled van der Pol oscillators," IEEE Trans. Circuits Syst., vol.25, no.1, pp.7–18, Jan. 1978.
- [9] S. Moro, Y. Nishio, and S. Mori, "Synchronization phenomena in oscillators coupled by one resister," IEICE Trans. Fundamentals, vol.E78-A, no.2, pp.244–253, Feb. 1995.
- [10] Y. Setou, Y. Inami, Y. Nishio, and A. Ushida, "On a ladder of oscillators sharing inductors," Proc. NOLTA'97, vol.1, pp.565–568, Nov. 1997.
- [11] M. Yamauchi, M. Wada, Y. Nishio, and A. Ushida, "Wave propagation phenomena of phase states in oscillators coupled by inductors as a ladder," IEICE Trans. Fundamentals, vol.E82-A, no.11, pp.2592–2598, Nov. 1999.
- [12] M. Okuda, M. Yamauchi, Y. Nishio, and A. Ushida, "Analysis of wave-motion in coupled oscillators by simplified model," Proc. NDES'00, pp.105–108, May 2000.
- [13] M. Yamauchi, M. Wada, Y. Nishio, and A. Ushida, "Collisions between two phase-inversion-waves in an array of oscillators," Proc. ISCAS'00, vol.1, pp.679–682, May 2000.
- [14] M. Yamauchi, Y. Nishio, and A. Ushida, "Reflection and transmission of phase-inversion-waves in oscillators coupled by two kinds inductors," Proc. ISCAS'01, vol.3, pp.767– 770, May 2001.



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